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Grid-connected Inverter Control Strategy Based on Capacitor Voltage Error Inertia Feedback

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Abstract. Firstly, the mathematical model of single-phase grid connected inverter is established. Then, the mathematical derivation is carried out accord to the state space model, and the control strategy of grid-connected inverter based on capacitor voltage error inertia feedback is proposed. In this strategy, the first-order inertial feedback of capacitor voltage error is adopted to suppress the resonance peak of LCL filter, the feedforward of capacitor reference voltage is adopted to improve the tracking performance of the system, and the proportional feedforward of grid-connected voltage is used to suppress the harmonic of power grid. In addition, Simulation verify the feasibility and effectiveness of the proposed control strategy.

1. Introduction

Renewable energy, represented by solar energy and wind energy, has become the focus of energy development in various countries due to its abundant reserves and no pollution. Renewable energy is generate electricity through distributed generation system. The principle is to change renewable energy sources into direct current power through power generation equipment, then the grid-connected inverter is used to change direct current (DC) power to alternating current (AC) power and transmit it to the grid. Therefore, grid connected inverter is core device of whole distributed generation system. In order to enable the distributed generation system to deliver high quality current to the power grid stably and efficiently, a good control strategy is needed to control the grid-connected inverter. Therefore, it has very important meaning to study the strategy of grid-connected inverter.

At present, for the sake of reducing high frequency harmonic components in the grid-connected current and obtain high-quality output current, LCL filter is usually selected to filter out harmonics[1]. However, LCL filter is a high-order system. Its high-order characteristics will make the system produce undamped resonance, and reduce the stability of the system. In addition, PI controllers[2]-[3] and PR controllers[4] are commonly used at present. However, PI control will have some tracking errors when tracking the AC signal, the AC signal can be converted to the DC signal through coordinate transformation to overcome the shortcoming that PI controller cannot track AC signal accurately. However, the application of this technology in single-phase inverter not only makes the control strategy for grid-connected inverter based on capacitive voltage error inertial feedback is proposed. The proposed control strategy is not simply to use the filter capacitor voltage and inverter side current for direct feedback. Instead, the error of the filter capacitance current and its reference

Content from this work may be used under the terms of the Creative Commons Attribution 3.0 licence. Any further distribution of this work must maintain attribution to the author(s) and the title of the work, journal citation and DOI. Published under licence by IOP Publishing Ltd 1 current is combined with the first order inertial controller as the inner loop feedback, and the error of the invert side current and its reference current is combined with the proportional controller as the feedback of the outer loop, through this to improve the robustness of the system. The proposed control strategy makes the grid-connected system have good grid-connected current quality, good tracking performance, good dynamic performance .At the same time, it has strong robustness to the fluctuation of grid voltage.

2. Mathematical Derivation of the Control Strategy

Figure1 is structure diagram of grid-connected inverter. Because the parasitic resistances of the inductor and capacitor is beneficial to the stability of system, and these resistances are relatively small. Therefore, in order to simulate the worst operating conditions of the system, the parasitic resistance of capacitance and inductance was ignored in this study[5].



Figure 1.structure diagram of inverter

Figure 2.Control diagram of LCL

 C_d is the DC-side capacitor, L_1 is the inverter-side inductor, i_1 is the inverter-side current, i_c is the capacitor current, C is the filter capacitor, L_2 is the grid-side inductor, u_g is the grid voltage, u_d is the DC-side voltage, u_c is the capacitor voltage, u_i is output voltage of inverter, i_2 is the grid-connected current.

According to figure 1, the LCL model structure figure 2 can be obtained. According to figure 2, the transfer function expression from u_i to i_2 in the LCL control diagram can be geted as follows:

$$\frac{I_2(s)}{U_i(s)} = \frac{1}{L_1 L_2 C s^3 + (L_1 + L_2) s}$$
(1)

According to figure 1, the differential equation of the inverter can be written as equation (2).

$$\begin{cases} \frac{di_{1}}{dt} = \frac{1}{L_{1}}u_{1} - \frac{1}{L_{1}}u_{c} \\ \frac{du_{c}}{dt} = \frac{i_{c}}{C} \\ \frac{di_{2}}{dt} = \frac{u_{c}}{L_{2}} - \frac{u_{g}}{L_{2}} \end{cases}$$
(2)

Where $u_i = \lambda u_d$, λ is the switch function, u_i is the control input of the inverter.

In this control strategy, the error of the filter capacitance current and its reference current is combined with the first order inertial controller as the inner loop feedback, and the error of the invert side current and its reference current is combined with the proportional controller as the feedback of the outer loop. So, define the state variables as (3).

$$x_1 = \dot{i}_1 - \dot{i}_{1r}$$
, $x_2 = u_c - u_{cr}$ (3)

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Where i_{1r} are the reference current of i_1 , and u_{cr} is the reference voltage of u_c . Making use of equation (3) can be get(4).

$$\dot{x}_{1} = \dot{i}_{1} - \dot{i}_{1r} = \dot{i}_{1} - \dot{i}_{cr} - \dot{i}_{2r}$$
⁽⁴⁾

Substitute the first formula in (2) into (4) to get(5).

$$\dot{x}_{1} = \frac{1}{L_{1}} u_{i} - \frac{1}{L_{1}} u_{c} - \dot{i}_{cr} - \dot{i}_{2r}$$
⁽⁵⁾

Making use of $u_c = x_2 + u_{cr}$, equation (5) can be written as equation (6).

$$\dot{x}_{1} = \frac{1}{L_{1}} u_{i} - \frac{1}{L_{1}} x_{2} - \frac{1}{L_{1}} u_{cr} - \dot{i}_{cr} - \dot{i}_{cr}$$
(6)

In order to minimize the error energy of the system, we only need to set the right side of equation (6) as 0, so it can get the following.

$$u_{i} = x_{2} + u_{cr} + L_{1}\dot{i}_{2r} + L_{1}\dot{i}_{cr}$$
⁽⁷⁾

By introducing the intermediate input variable u_{i1} into equation (7), equation (7) can be written as equation (8).

$$u_{\rm i} = u_{\rm i1} + u_{\rm cr} + L_{\rm i}\dot{i}_{\rm 2r} + L_{\rm i}\dot{i}_{\rm cr}$$
(8)

In order to accelerate x_1 convergence, u_{i1} introduces the state feedback gain k_1 , u_{i1} can be written as equation (9).

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$$u_{i1} = u_{i2} - k_1 x_1 \tag{9}$$

 u_{i2} introduces the state feedback gain k_2 , u_{i2} can be written as equation (10).

$$u_{i2} = -k_2 x_2 \tag{10}$$

Substitute equation(10) and (9) into (8), we can get (11).

$$u_{i} = -k_{1}x_{1} - k_{2}x_{2} + u_{cr} + L_{1}\dot{i}_{2r} + L_{1}\dot{i}_{cr}$$
(11)

Then, making use of $i_{cr} = C(du_{cr}/dt)$ and $u_i = \lambda u_d$, the control duty cycle can be obtained as (12).

$$\lambda = \frac{1}{u_{d}} \left[-k_{1}x_{1} - k_{2}x_{2} + u_{cr} + L_{1}C\frac{d^{2}u_{cr}}{dt^{2}} + L_{1}\frac{di_{2r}}{dt} \right]$$
(12)

In this design, PI controller is used to generate the reference current i_{1r} of current i_1 . The transfer function of PI is shown in (13).

$$G_{\rm PI}(s) = k_{\rm p} + k_{\rm i} / s \tag{13}$$

According to formula (12), the control structure diagram 3 of the system can be obtained.



Figure 3. Control structure of grid-connected system

It can be know from Figure.3 that the second order differential is contained in the capacitor reference voltage feedforward control. In practical engineering applications, the differential link is easier to introduce noise[6]. In this study, low-pass filter $1/(\tau s + 1)$ is introduced into the capacitor reference voltage feedforward in order to suppression noise, where τ is time constant. Moreover, due to the large number of harmonics in the power grid, they will be harmful to the quality of grid-connected current, and feedforward of grid-connected voltage is widely used due to the good effect of suppressing the harmonics. Therefore, this paper adopts grid-connected voltage proportional feedforward to suppress power grid harmonics. At the same time, in order to suppress the resonance of LCL filter, the first-order inertia link $1/(\tau s + 1)$ is introduced into the capacitor voltage error feedback[7].

So, the improved grid-connected inverter system control duty cyclecan can be written as equation (14).

$$\lambda = \frac{1}{u_{\rm d}} \left[-k_{\rm l} x_{\rm l} - k_{\rm 2} \tau^{-1} e^{-t/\tau} x_{\rm 2} + \tau^{-1} e^{-t/\tau} (u_{\rm cr} + L_{\rm l} C \frac{d^2 u_{\rm cr}}{dt^2}) + k u_g \right]$$
(14)

Where $\tau^{-1}e^{-t/\tau}$ is the inverse Laplace transform of $1/(\tau s+1)$.

According to formula (14), the mproved control structure diagram 4 of the system can be obtained.



Figure 4. improved control structure of the grid-connected system

Where $u_{cr} = L_2 s i_{2r} + u_g$, PWM is the inverter link, $G_h(s) = [1/(\tau s + 1)](CL_1 s^2 + 1)$, $G_c(s) = k_2 [1/(\tau s + 1)]$.

It is obvious that the system control structure in figure 4 is mainly composed of five controllers . 1) ku_g is the grid-connected voltage proportional feedforward controller, which can suppress the

influence of harmonics on the system.

2) $[1/(\tau s+1)](CL_1s^2+1)u_{cr}$ is the capacitor reference voltage feedforward controller. It can reduce the phase difference between i_2 and u_g , improve the tracking performance of the system.

3) $k_2[1/(\tau s+1)]x_2$ is the first-order inertial feedback controller of capacitance voltage error, which can provide some active damping for the system, and can suppress the resonance peak of LCL filter.

4) k_1x_1 is the traditional P controller.

5) $G_{\text{PI}}(s)(i_{2r}-i_2)$ is the PI controller to generate reference current of inverter side, which can improve the performance of system.

3. The influence of inertial feedback on system resonance

By substituting equation (14) into equation (2), and then carrying out Laplace transform and simplification, the closed-loop transfer function between i2 and its reference current i2r of the system can be obtained as (15).

$$\frac{I_2}{I_{2r}} = \frac{A_4 s^4 + A_2 s^2 + A_1 s + A_0}{B_5 s^5 + B_4 s^4 + B_3 s^3 + B_2 s^2 + B_1 s + B_0}$$
(15)

$$B_{5}=L_{2}L_{1}C\tau, B_{4}=L_{2}(L_{1}C+k_{1}C\tau), B_{3}=L_{1}\tau+k_{1}CL_{2}+\tau L_{2}, B_{2}=k_{1}\tau+L_{1}+k_{1}\tau k_{p}+k_{2}L_{2}+L_{2}, B_{1}=k_{1}+k_{1}k_{p}+k_{1}k_{i}\tau, B_{0}=k_{1}k_{i}, A_{4}=L_{1}CL_{2}, A_{2}=k_{1}\tau k_{p}+k_{2}L_{2}+L_{2}, A_{1}=k_{1}k_{p}+k_{1}\tau k_{i}, A_{0}=k_{1}k_{i}$$

Then, the open-loop transfer function (16) of system can be obtained.

$$G_{1} = \frac{I_{2}(s)}{I_{2r}(s) - I_{2}(s)} = \frac{A_{4}s^{4} + A_{2}s^{2} + A_{1}s + A_{0}}{C_{5}s^{5} + C_{4}s^{4} + C_{3}s^{3} + C_{2}s^{2} + C_{1}s + C_{0}}$$
(16)

Where $C_5 = B_5$, $C_4 = B_4 - A_4$, $C_3 = B_3$, $C_2 = B_2 - A_2$, $C_1 = B_1 - A_1$, $C_0 = B_0 - A_0$.

According to equations (1) and (16), the bode diagram of transfer function from u_i to i_2 in LCL model and the bode diagram of the system's open-loop can be drawn respectively, this is shown in Figure 5. As can be seen from the figure, the bode diagram curve of LCL model has a resonance peak. However, the bode diagram curve of the system designed has no obvious resonance peak.

4. Simulation and result analysis

In order to make further verify the effectiveness and feasibility of the control strategy proposed in this paper, simulink is used for simulation[8]. The simulation model is built in Simulink according to the

Figure 4. The simulation results are shown in Figure 6-9. Set the given value of grid-connected current as $i_{2i}=15$ sin(wt), the effective value of grid voltage is 220V. In order to ensure a good visual effect and facilitate observation, the grid-connected voltage is reduced to its 0.2 times before entering the oscilloscope. The parameters as: fundamental frequency $f_0=50$ Hz, switching frequency $f_{sw}=20000$ Hz, $u_d=400$ V, $L_1=2$ mH, C=12uF, $L_2=0.8$ mH, $k_1=170$, $k_2=80$, k=1, $k_i=18000$, $k_p=1$, w=314rad/s, $\tau=0.01$ ms

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Figure 5. Bode diagram of transfer function of LCL model and

open-loop transfer function of grid connected system

Figure 6 shows the simulation oscillograms of the grid-connection voltage u_g and grid-connected current i_2 and the fast Fourier transform(FFT) analysis diagram of grid-connected current i_2 for the system. In figure 6, the simulation waveforms of u_g and i_2 are all relatively smooth sinusoidal waveforms. The total harmonic distortion rate (THD) of i_2 is only 0.40%, which is far less than the international standard 5%. The simulation peak value of grid-connected current i_2 is about 15.05, almost equal to the given value 15. According to the simulation in figure 6, the proposed control strategy enables the system to have good grid-connected current quality and good tracking performance.



Figure 6. The simulation oscillograms of u_g and i_2 for the system

Figure 7 is simulation oscillograms of u_g and i_2 for the system with sudden change of reference current setting value. Where, (a) is the overall simulation oscillograms of u_g and i_2 of the reference current given value mutation, and (b) is the local simulation oscillograms of u_g and i_2 of the reference current given value mutation. That can know from (a), when reference current i_{2r} decreases from 15A to 7.5A at t=0.045s, the simulation curves of u_g and i_2 can track the change of reference current i_{2r} . As can be seen from figure (b), When the grid-connected reference current i_{2r} changes, the grid-connected current quickly tracks the reference current a after experiencing a short fluctuation of 0.7ms. The above analysis shows that the system has better dynamic performance.



Figure 7. the simulation waveforms of u_g and i_2 with sudden

change of reference current setting value.

Figure 8 is simulation oscillogramsa of u_g and i_2 and FFT diagram of i_2 for the system when the grid voltage surges. Where (a) is the simulation oscillogram and (b) is the FFT diagram. That can know from (a), the grid voltage increases by 20% at t=0.06s, that is, it increases from the original 311V to 373.2V, the grid-connected current can remain stable. In addition, that can know from (b) that the THD of the system is 0.45% when the grid voltage is increased by 20%, which is far less than the international standard of 5%. These show that the system has strong robustness to the fluctuation of power grid voltage.



Figure 8.the simulation oscillograms of u_g and i_2 and FFT diagram of i_2

When the grid voltage increases by 20%

Figure 9 show that the simulation oscillograms of u_g and i_2 of the system under with and without capacitor reference voltage feedforward, respectively. As can be seen from Fig 9, the phase difference of u_g and i_2 with capacitor reference voltage feedforward is smaller than that of u_g and i_2 without capacitor reference voltage feedforward. This shows that the capacitor reference voltage feedforward is helpful to improve the tracking accuracy of the system. The grid power of system connection is improved.

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Figure9 the simulation oscillograms of u_g and i_2 of the system

5. Conclusion

The mathematical derivation is carried out accord to the state space model of the system the control strategy of grid-connected inverter based on capacitance voltage error inertia feedback is obtained. The control strategy consists of five controllers, which respectively the first-order inertial feedback controller of capacitor voltage error to suppress resonance, and the capacitor reference voltage feedforward controller to improve the tracking performance of the system, and the grid-connected voltage proportional feedforward controller to to suppress the influence of grid harmonics on the system, and the PI controller to generate reference current of inverter-side current, and the traditional P controller. Simulation results show that the proposed control strategy makes the grid-connected system have good grid-connected current quality, good tracking performance, good dynamic performance .At the same time, it has strong robustness to the fluctuation of grid voltage.

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